## SEEU2012

Electronics
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## Chapter 4 Bipolar Junction Transistor (BJT) AC Analysis

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## Course Learning Outcomes

Apply the basic law and theorems of electronic devices to describe their basic operation.

Apply the basic law, theorems and methods of analysis to solve complex problem related to circuitry.

Work in a team and communicate effectively.

## Small Signal Circuit Analysis

to determine the initial operating values of $I_{C}, I_{B}$ and $V_{C E}(Q-$ point). The Q-point which is in the middle of the DC load line is

to determine the values of input impedance $\left(Z_{i}\right)$, output impedance $\left(Z_{0}\right)$, voltage gain $\left(A_{v}\right)$ and current gain $\left(A_{i}\right)$

For the purpose of analysing AC operations, the transistor can be replaced with a small signal equivalent circuit model when it is operating in the active region (having linear attributes)


AC Analysis


## Function Of Capacitors In Amplifiers



## DC Analysis

$\square$ Set AC source to zero.
Replace the coupling capacitors and bypass capacitor with open circuits.


## AC Analysis

$\square$ Set DC source to zero (ground)
$\square$ Replace the coupling capacitors and bypass capacitors with short circuit
$\square$ Analyze the AC operation


## AC Equivalent Models

$\square$ Linear transistor can be replaced by AC equivalent model.
$\square$ Transistor AC equivalent models are:

- Parameter-h Model
- $r_{e}$ Model
- Parameter-y Model
- Hybrid- $\pi$ Model


## Transistor AC Models

The simplified Hybrid- $\pi$ are shown in relation to the transistor model

An important Hybrid $-\pi$ is $r_{\pi}$. It appears as a small AC resistance between the Base and Emitter


## Hybrid - $\pi$ Model



$$
\beta=\mathrm{h}_{\mathrm{fe}} \quad r_{o}=\frac{1}{h_{o e}}=\frac{V_{A}}{I_{C}} \quad g_{m}=\frac{I_{C}}{V_{T}} \quad r_{\pi}=\frac{\beta}{g_{m}}
$$

$\square$ In the Current Controlled Current Source (CCCS) model $=\beta \mathrm{i}_{\mathrm{b}}$
$\square$ In the Voltage Controlled Current Source (VCCS) model $=g_{m} v_{\pi}$
Internal output impedance, $r_{o}$ is very high \& can be neglected(open circuit)Equivalent voltage temperature, $\mathrm{V}_{\mathrm{T}}=26 \mathrm{mV}$ at room temperature $\left(25^{\circ} \mathrm{C}\right)$
$\square$ EarlyVoltage, $\mathrm{V}_{\mathrm{A}} \cong 200 \mathrm{~V}$

## Hybrid - $\pi$ Model



## Hybrid - $\pi$ Model



EQUATION TO REMEMBER!

$$
\beta=\mathrm{h}_{\mathrm{fe}} \quad r_{o}=\frac{1}{h_{o e}}=\frac{V_{A}}{I_{C}} \quad g_{m}=\frac{I_{C}}{V_{T}} \quad r_{\pi}=\frac{\beta}{g_{m}}
$$

## Transistor Configuration

$\square$ Transistor configuration - is a connection of transistor to get variety operation.
$\square 3$ types of transistor configuration:

- Common Collector (CC).
- Common Base (CB).
- Common Emitter (CE).
$\square$ Common means the circuit has a single reference for both the input voltage to the transistor and the output voltage


|  | Common <br> Emitter | Common <br> Collector | Common Base |
| :---: | :---: | :---: | :---: |
| Input terminal | Base | Base | Emitter |
| Output Terminal | Collector | Emitter | Collector |
| Common (Gnd) | Emitter | Collector | Base |
|  |  |  |  |

## Transistor Configuration

## Example

## Determine the configuration of the following BJT circuit?



## Exercise

Determine the configuration of the following BJT circuit?


## Exercise

Determine the configuration of the following BJT circuit?

(D)



## Common Emitter

$\square$ Input-Applied to BASE
$\square$ Output - From COLLECTOR
$\square$ High Voltage Gain, $A_{V}$ and high Current Gain, $A_{i}$
$\square$ Phase shift between input and output is $180^{\circ}$


# Common Emitter: Fixed Bias Without By Pass Capacitor 



CE Amplifier


DC source grounded and capacitors are shorted
$\square$ Set all DC source to zero/ground
$\square$ Replace the coupling capacitors and bypass capacitors with short circuit
$\square$ Rearrange circuit to make it simple and draw the AC equivalent circuit.

## Common Emitter: Fixed Bias



## Common Emitter : Fixed Bias

Voltage gain:

$$
\begin{aligned}
& A_{v s}=\frac{v_{o}}{v_{s}} \\
& v_{o}=-g_{m} v_{\pi}\left(R_{C} \| R_{L}\right) \\
& v_{\pi}=v_{i}=\frac{Z_{i}}{Z_{i}+R_{S}} v_{s} \\
& A=\frac{v_{o}}{v_{\pi}} \times \frac{v_{\pi}}{v_{i}} \times \frac{v_{i}}{v_{s}}
\end{aligned}
$$

$$
A_{v s}=-g_{m}\left(R_{C} \| R_{L}\right) \frac{Z_{i}}{Z_{i}+R_{S}}
$$

## Current gain:

$$
A_{i}=\frac{i_{o}}{i_{i}}
$$

$$
i_{o}=\frac{R_{C}}{R_{C}+R_{L}}\left(-g_{m} v_{\pi}\right)
$$

$$
v_{\pi}=v_{i}=i_{i} Z_{i}
$$

$$
A_{i}=\frac{i_{o}}{v_{\pi}} \times \frac{v_{\pi}}{v_{i}} \times \frac{v_{i}}{i_{i}}
$$

$$
A_{i}=-\frac{R_{C}}{R_{C}+R_{L}} g_{m} Z_{i}
$$

## Common Emitter : Fixed Bias (Exercise)



## Common Emitter : Fixed Bias (Solution)

1) Draw the DC equivalent circuit - ground all DC supply and short all capacitors.
2) Draw the AC equivalent circuit


## Common Emitter: Voltage Divider Bias

$\square$ Voltage Divider Bias


## Common Emitter: Voltage Divider Bias



## Exercise

Draw the AC equivalent circuit for the following voltage divider bias circuit:


## Common Emitter: Voltage Divider Bias Example

Determine the values of $Z_{\mathrm{i}}, Z_{\mathrm{o}}, A_{\mathrm{v}(\mathrm{oc})}, A_{\mathrm{v}}, A_{\mathrm{vs},}, A_{\mathrm{i}}$ and $A_{\text {is }}$ for the amplifier circuit below. Given $V_{A}=200 \mathrm{~V}$ and $I_{C}=2.37 \mathrm{~mA}$


## Solution



Input impedance: $\quad Z_{\mathrm{i}}=R_{1}\left\|R_{2}\right\| r_{\pi}=8.2 \mathrm{k} \Omega|1.5 \mathrm{k} \Omega| 1.32 \mathrm{k} \Omega=647 \Omega$
Output impedance: $\quad Z_{\mathrm{o}}=r_{\mathrm{o}}\left\|R_{\mathrm{C}}=84.4 \mathrm{k} \Omega\right\| 4.3 \mathrm{k} \Omega=4.09 \mathrm{k} \Omega$
Voltage gain: $\quad A_{v(\text { oc })}=\frac{v_{\text {out }}}{v_{\text {in }}}=-g_{\mathrm{m}}\left(r_{\mathrm{o}} \| R_{\mathrm{C}}\right)=-(91.2 \mathrm{mS})(4.09 \mathrm{k} \Omega)=-373$

$$
A_{v}=\frac{v_{\text {out }}}{v_{\text {in }}}=A_{\text {voc) }}\left(\frac{R_{\mathrm{L}}}{Z_{\mathrm{o}}+R_{\mathrm{L}}}\right)=(-373)\left(\frac{10 \mathrm{k} \Omega}{4.09 \mathrm{k} \Omega+10 \mathrm{k} \Omega}\right)=-265
$$

$$
A_{\text {vs }}=\frac{v_{\text {out }}}{v_{\mathrm{s}}}=A_{v}\left(\frac{Z_{\mathrm{i}}}{R_{\mathrm{s}}+Z_{\mathrm{i}}}\right)=(-265)\left(\frac{647 \Omega}{600 \Omega+647 \Omega}\right)=-137.5
$$

## Current gain:

$$
\begin{gathered}
A_{\mathrm{i}}=\frac{i_{\text {out }}}{i_{\text {in }}}=A_{\mathrm{v}(\mathrm{oc})}\left(\frac{Z_{\mathrm{i}}}{Z_{\mathrm{o}}+R_{\mathrm{L}}}\right)=(-373)\left(\frac{647 \Omega}{4.09 \mathrm{k} \Omega+10 \mathrm{k} \Omega}\right)=-17.13 \\
A_{\mathrm{is}}=\frac{i_{\text {out }}}{i_{\mathrm{s}}}=A\left(\frac{R_{\mathrm{s}}}{R_{\mathrm{s}}+Z_{\mathrm{i}}}\right)=(-17.13)\left(\frac{600 \Omega}{600 \Omega+647 \Omega}\right)=-8.24
\end{gathered}
$$

## Common Emitter: Unbypassed $\mathrm{R}_{\mathrm{E}}$



The removal of the bypass capacitor results in:
$\square$ an increase in the input impedance, $Z_{i}$ \& output impedance, $Z_{\text {o }}$
$\square$ a reduction in its voltage gain, $A_{v}$

## Common Emitter: Unbypassed $\mathrm{R}_{\mathrm{E}}$

Input impedance:
$Z_{\mathrm{i}}=R_{\mathrm{B}} \| Z_{\mathrm{b}}$
$Z_{\mathrm{b}}=r_{\pi}+(\beta+1) R_{\mathrm{E}}$
Output impedance (ignore $r_{0}$ ): $Z_{o}=R_{C}$

Voltage gain:

$$
\begin{aligned}
& A_{v(0 c)}=\frac{v_{0}}{v_{\mathrm{i}}}=-\left.\frac{g_{\mathrm{m}} v_{\pi} R_{\mathrm{C}}}{i_{\mathrm{b}} Z_{\mathrm{b}}}\right|_{z_{\mathrm{b}}=r_{\mathrm{a}}+(\beta+1) R_{\mathrm{E}}} \\
& =-\frac{g_{\mathrm{m}} i_{\mathrm{b}} r_{\pi} R_{\mathrm{C}}}{i_{\mathrm{b}}\left(r_{\pi}+g_{\mathrm{m}} r_{\pi} R_{\mathrm{E}}\right)}=-\frac{g_{\mathrm{m}} R_{\mathrm{C}}}{1+g_{\mathrm{m}} R_{\mathrm{E}}} \\
& A_{\mathrm{V}(0 \mathrm{C})}=\frac{v_{\mathrm{o}}}{v_{\mathrm{i}}} \cong-\left.\frac{R_{\mathrm{C}}}{R_{\mathrm{E}}}\right|_{g_{\mathrm{m}} R_{\mathrm{E}} \gg 1}
\end{aligned}
$$



Current gain:

$$
A_{i}=\frac{i_{o}}{i_{i}}=-\frac{g_{m} r_{\pi} R_{B}}{R_{B}+Z_{b}}=-\frac{\beta R_{B}}{R_{B}+Z_{b}}
$$

Current gain from voltage gain:

$$
A_{i}=A_{v(c)}\left(\frac{Z_{i}}{Z_{o}+R_{L}}\right)=A_{v(c)}\left(\frac{Z_{i}}{R_{c}}\right)
$$

## Common Emitter: Unbypassed $\mathrm{R}_{\mathrm{E}}$

To determine $Z_{o}$ using imaginary voltage source (alternative approach)


$$
\begin{aligned}
Z_{\text {TEST }} & =\frac{v_{\text {TEST }}}{i_{\text {TEST }}} \\
v_{\text {TEST }} & \approx\left(i_{\text {TEST }}-g_{\mathrm{m}} v_{\pi}\right) r_{\mathrm{O}}+i_{\text {TEST }} R_{\mathrm{E}} \\
& \approx i_{\text {TEST }} r_{\mathrm{O}}-g_{\mathrm{m}} v_{\pi} r_{\mathrm{O}}+i_{\text {TEST }} R_{\mathrm{E}}
\end{aligned}
$$

## Common Emitter: Unbypassed $\mathrm{R}_{\mathrm{E}}$

$$
\begin{aligned}
i & =\frac{R_{\mathrm{E}}}{R_{\mathrm{E}}+\left(r_{\pi}+R_{\text {TH }}\right)} \times i_{\text {TEST }} \\
v_{\pi} & =-i r_{\pi}=\frac{-R_{\mathrm{E}} r_{\pi}}{R_{\mathrm{E}}+r_{\pi}+R_{\text {TH }}} \times i_{\text {TEST }} \\
v_{\text {TEST }} & =i_{\text {TEST }} r_{\mathrm{O}}-g_{\mathrm{m}} r_{\mathrm{O}}\left[\frac{-R_{\mathrm{E}} r_{\pi}}{R_{\mathrm{E}}+r_{\pi}+R_{\text {TH }}}\right] \times i_{\text {TEST }}+i_{\text {TEST }} R_{\mathrm{E}} \\
& =i_{\text {TEST }}\left[r_{\mathrm{O}}+\frac{g_{\mathrm{m}} r_{\mathrm{O}} R_{\mathrm{E}} r_{\pi}}{R_{\mathrm{E}}+r_{\pi}+R_{\text {TH' }}}+R_{\mathrm{E}}\right] \\
& \therefore Z_{\text {TEST }}=\frac{v_{\text {TEST }}}{i_{\text {TEST }}}=r_{\mathrm{O}}+\frac{g_{\mathrm{m}} r_{0} R_{\mathrm{E}} r_{\pi}}{R_{\mathrm{E}}+r_{\pi}+R_{\text {TH }}}+R_{\mathrm{E}} \\
& \therefore Z_{\mathrm{O}}=\left(Z_{\text {TEST }} \| R_{\mathrm{C}}\right) \approx R_{\mathrm{C}}
\end{aligned}
$$

## Example of Common Emitter: Unbypassed $\mathrm{R}_{\mathrm{E}}$

Determine the values of $Z_{\mathrm{i}}, Z_{\mathrm{o}^{\prime}}, A_{\mathrm{v}(\mathrm{oc})}, A_{\mathrm{v}}, A_{\mathrm{vs}}, A_{\mathrm{i}}, A_{\mathrm{is}}$ and $A_{\mathrm{p}}$ for the amplifier circuit shown below. Given $V_{A}=\infty$ and $I_{C}=2.37 \mathrm{~mA}$


## Example of Common Emitter: Unbypassed $\mathrm{R}_{\mathrm{E}}$

Input impedance:


$$
\begin{aligned}
& Z_{\mathrm{b}}=r_{\pi}+(\beta+1) R_{\mathrm{E}}=1.32 \mathrm{k} \Omega+(121)(1 \mathrm{k} \Omega)=122.32 \mathrm{k} \Omega \\
& \quad Z_{\mathrm{i}}=R_{1}\left\|R_{2}\right\| Z_{\mathrm{b}}=8.2 \mathrm{k} \Omega\|1.5 \mathrm{k} \Omega\| 122.32 \mathrm{k} \Omega=1.26 \mathrm{k} \Omega
\end{aligned}
$$

Output impedance: $\quad Z_{\mathrm{o}}=R_{\mathrm{C}}=4.3 \mathrm{k} \Omega$
Voltage gain: $A_{v(0 c)}=\frac{v_{\text {out }}}{v_{\text {in }}}=-\frac{g_{\mathrm{m}} R_{\mathrm{C}}}{1+g_{\mathrm{m}} R_{\mathrm{E}}}=-\frac{(91.2 \mathrm{mS})(4.09 \mathrm{k} \Omega)}{1+(91.2 \mathrm{mS})(1 \mathrm{k} \Omega)}=-4.25$

## Exercise

Draw the small signal ac equivalent circuit. If $+\mathrm{V}_{\mathrm{CC}}=12 \mathrm{~V}, \mathrm{~V}_{\mathrm{A}}=\infty, \mathrm{V}_{\mathrm{BE}}=$ 0.7 V and $\beta=80$, calculate the values of :
(a) Input and output impedances
(b) Voltage gain $A_{V}$ and $A_{v s}$


## Exercise

Given $\beta=h_{\text {fe }}=200, \mathrm{~V}_{\mathrm{BE}}=0.6 \mathrm{~V}, \mathrm{~V}_{\mathrm{CEQ}}$ $=10 \mathrm{~V}, \mathrm{~V}_{\mathrm{A}} \rightarrow \infty, \mathrm{A}_{\mathrm{v}}(\mathrm{dB})$ at middle frequency $=14 \mathrm{~dB}$.
(a) Draw the mid-frequency AC equivalent circuit.
(b) Determine $\mathrm{R}_{\mathrm{E} 1}$ and $\mathrm{R}_{\mathrm{E} 2}$.
(c) Determine the amplifier input impedance, $\mathrm{Z}_{\mathrm{i}}$.
(d) Ratio of $i_{L} / v_{i}$.


$$
\begin{aligned}
& R_{E 1}=56.67 \Omega ; R_{E 2}=7.27 \mathrm{k} \Omega \\
& Z_{i}=2.86 \mathrm{k} \Omega ; i_{L} / v_{i}=4.99 \mathrm{mS}
\end{aligned}
$$

